Highly Linear Ka-Band 0.15µm GaAs Power pHEMT Process for Use in Low-Cost Molded QFN Plastic Package

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Abstract

Next generation communication radio systems are designed to use highly linear but low-cost power amplifiers in Ka-band. To serve this demand, United Monolithic Semiconductors has developed a 0.15µm GaAs power pHEMT process offering outstanding transistor level linearity while maintaining state-of-the-art RF power density and gain performance up to 40GHz. This technology called PPH15X-20 can be used in low-cost molded QFN plastic package without any RF performance degradation. In this paper, we want to present the key figures of merit of this technology including two-tone on-wafer linearity performance.

INTRODUCTION

Today's market already offers power amplifier products in Ka-band with good linearity and gain performance to serve the demand for next generation communication radio systems [1,2]. However, these products have a major disadvantage. This is related to the need of robustness against humidity while maintaining the good RF performance. Therefore, these products need to be put in very expensive hermetic air cavity packages. This is necessary since the RF performance of the chip would be strongly degraded when adding a final chip protection layer. This chip protection layer is needed to protected the chip against humidity and to get compatible with cheap non-hermetic molded QFN plastic packages.

To fulfill all requirements in a low-cost molded QFN plastic package, good linearity, gain performance and robustness against humidity, UMS has developed a second generation of its 0.15µm Ka-band GaAs power pHEMT technology, called PPH15X-20.

TECHNOLOGY DESCRIPTION

The PPH15X-20 technology is utilizing a classical pseudomorphic AlGaAs/InGaAs/GaAs double recess HEMT structure. The 0.15µm T-gate is formed by a dielectric assisted gate process. It is designed for an operation voltage of V_{DD} = 6V. The typical electrical DC and RF device characteristics are summarized in table I.

TABLE I
TYPICAL DC AND RF DEVICE PARAMETERS OF PPH15X-20

DC Parameters (1x100µm Device)			
Parameter Name		**Value**	**Unit**
Maximum Drain Current	I_{DS+}	570	mA/mm
Drain Saturation Current	I_{DSS}	350	mA/mm
Maximum Transconductance	G_{max}	480	mS/mm
Threshold Voltage	$V_{g,100}$	-0.95	V
Three-Terminal Breakdown	V_{BDS}	≥12	V
Gate-Drain Diode Breakdown	V_{BGD}	≥-13	V
Small-Signal RF Parameters (2x75µm Device)			
Parameter Name		**Value**	**Unit**
Input Capacitance	C_{in}	130	fF
Feedback Capacitance	C_f	15	fF
Output Resistance	R_{out}	400	Ω
Extrinsic RF Transconductance	G_{me}	60	mS

In comparison to the first generation technology called PPH15X-10 [3], the gate module has been modified. A big part of the SiN used for device passivation has been replaced by a low dielectric material around the gate. This modification leads to a significantly reduced feedback capacitance C_f, reduces the influence of higher level SiN layers on the small-signal capacitances and supports the robustness of the device against humidity. Besides these positive effects, this kind of gate passivation eliminates the negative impact of the final chip protection on the intrinsic capacitances of the device which typically causes a degradation of the overall RF performance of the device.

The process offers, furthermore, two interconnect metal layers, where one can be optionally re-enforced for higher current density capability, as well as two thin film resistor layers with 30Ω□ and 1000Ω□, respectively. Furthermore, two capacitor densities with 260pF/mm^2 and 625pF/mm^2 can be integrated. When chips are put into non-hermetic plastic or air-cavity packages, a final chip protection is offered. The backside process is a 70µm process with individual-through-wafer-source-vias.

RF DEVICE CHARACTERIZATION

Small-Signal Characterization:

To characterize the gain improvement achieved mainly by the reduction of the small-signal feedback capacitance, as described in the section before, wafers with and without final BCB protection have been characterized by S-parameter measurements up to 50GHz. The maximum gain MSG/MAG has been calculated from these S-parameters and plotted versus frequency in fig. 1.

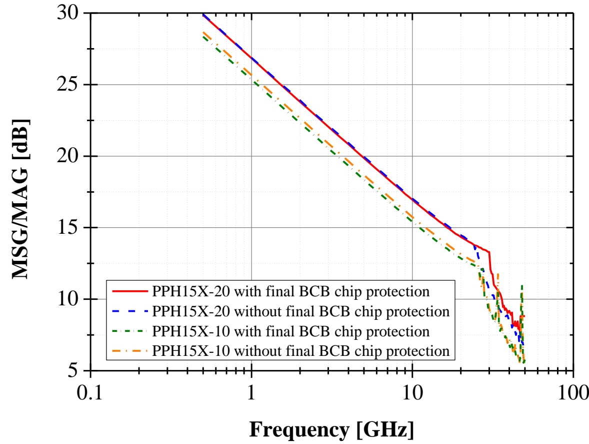

Fig. 1: MSG/MAG determined by S-parameter measurement up to 50GHz performed on a 8x75µm device at $V_{DS} = 6V$.

As can be seen, the small-signal RF gain is not impacted by the BCB protection. Whereas the first generation technology clearly shows a RF gain drop by approx. 0.5dB when a final BCB chip protection is applied. The graph, furthermore, shows the clear gain improvement by 1.5dB compared to the first generation technology due to the reduction of the feedback capacitance C_f.

Large-Signal Characterization:

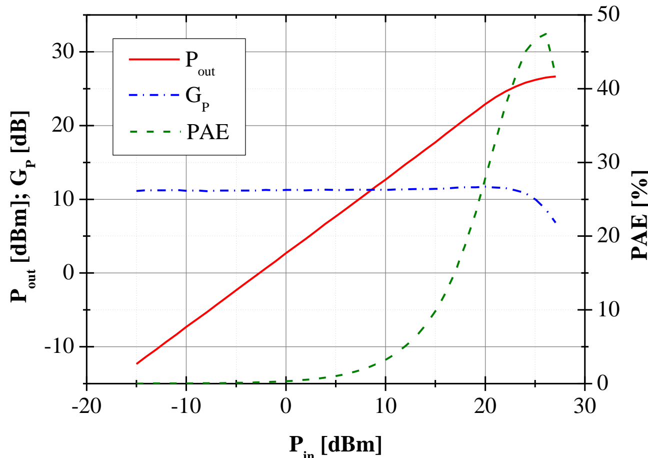

Fig. 2: Load-Pull power performance at 26GHz measured on a 8x75µm device at $V_{DS} = 6V$.

To assess the large-signal RF characteristics of the technology, on-wafer one-tone load-pull measurements have been carried out. These measurements have been performed on a representative 8x75µm device in µ-strip configuration at the desired operation voltage of $V_{DS} = 6V$.

The on-wafer one-tone load-pull measurements performed at 26GHz depicted in fig. 2 give a power density at the 1-dB-compression point of $P_{-1dB} = 750mW/mm$. The linear power gain is $G_p = 11dB$ and the peak Power Added Efficiency is $\eta_{PAE} = 48\%$.

Two-Tone Linearity Characterization:

The on-wafer two-tone load-pull measurements have been performed at 10GHz with a tone spacing of 10MHz. To determine the two-tone linearity performance represented by the third-order intermodulation distance IMD3, the IMD3 is plotted versus the respective two-tone output power and the results are compared to the first generation technology as well as to two commercial technologies.

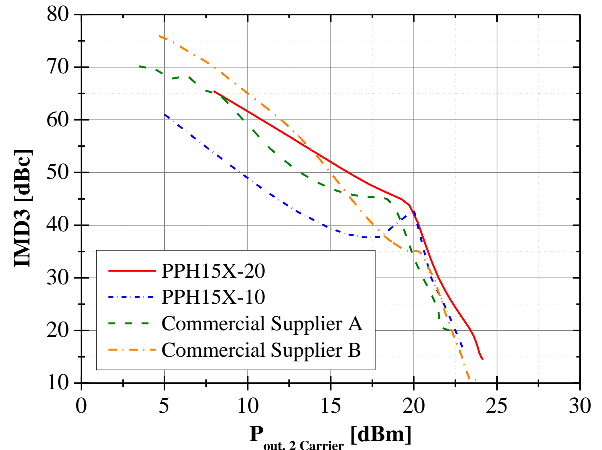

Fig. 3: Two-tone linearity performance versus output power at 10GHz and a tone spacing of 10MHz measured on a 8x75µm device at $V_{DS} = 6V$.

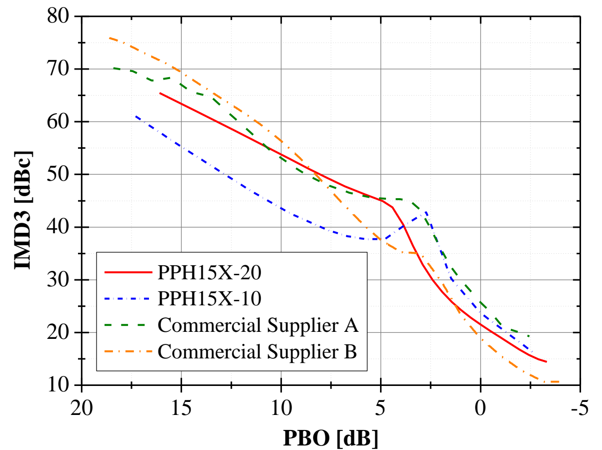

Fig. 4: Two-tone linearity performance versus power back-off at 10GHz and a tone spacing of 10MHz measured on a 8x75µm device at $V_{DS} = 6V$.

The results are depicted in fig. 3. It can bee seen that PPH15X-20 is presenting a very linear IMD3 performance up to the output power compression level without any sweet-spot formation. Especially in the range of 6…8dB PBO towards P_{-1dB} the technology is even showing slightly better linearity performance than all three compared technologies, as can be seen in fig.4 where the IMD3 is plotted versus PBO. This area is of special interest for the design of highly linear PAs for communication applications since the designs are optimized in this PBO region to achieve good linearity at a reasonably high output power level.
To investigate the impact of the final BCB chip protection layer on the linearity performance, devices with and without final BCB protection have been characterized by two-tone load-pull measurements and compared. The two-tone linearity results shown in fig. 5 prove that the final BCB chip protection does not degrade the linearity perfomance either. The small difference observed here is well within the the measurement accuracy of a two-tone load-pull system.

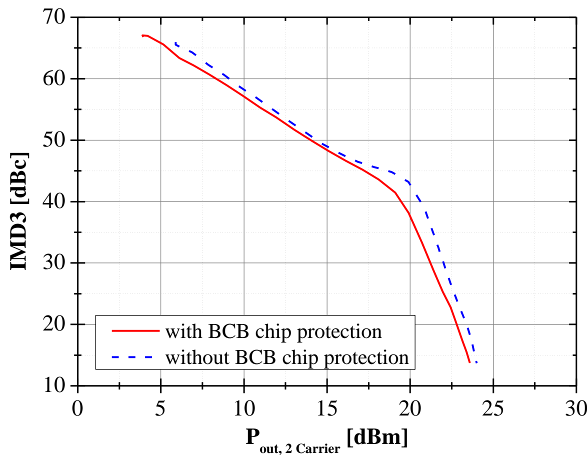

Fig. 5: Two-tone linearity performance at 10GHz and a tone spacing of 10MHz measured on a 8x75µm device at $V_{DS} = 6V$.

This investigation highlights the fact that the intrisic RF performance of the device is not impacted in any way by applying a final BCB chip protection. To this end, it can be stated, that the performance in QFN package is also unchanged when the chip is overmolded.

Reliability Assessment

Long-term accelerated life tests on single-stage 50Ω-matched amplifiers called DECs have been carried out to assess the reliability of the process. An example of RF gain and output power interim measurements during a HTOL test at a junction temperature of 270°C is plotted in fig. 6.
The output power degradation during HTOL tests is the limiting factor. The MTF of this test is $T_{50\%} = 1980h$ for a failure criterion of 1dB output power degradation. Based on preliminary life-tests at different junction temperatures a life time of more than 40 years has been extracted.

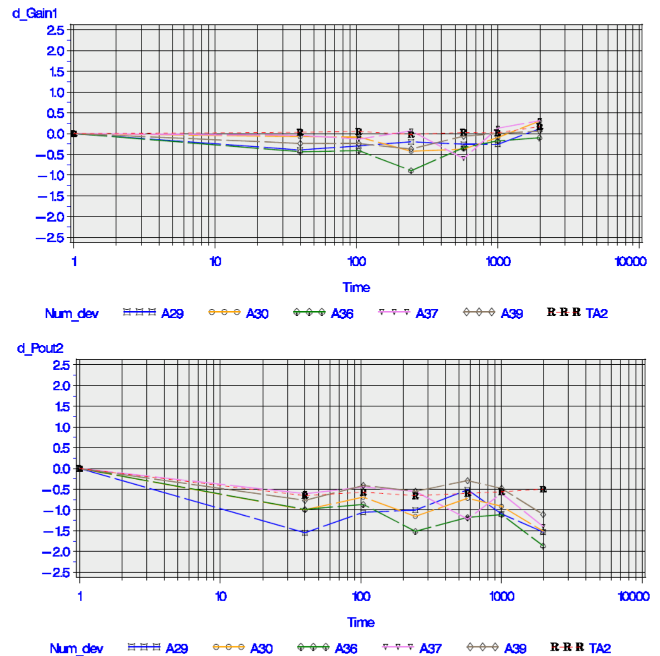

Fig. 6: RF gain and output power interim measurements of a HTOL test at a junction temperature of 270°C for 2000h.

To assess the robustness against humidity, full MMICs have been tested in an open package in biased temperature humidity tests (THB). The test conditions are a case temperature of $T_C = 85°C$ and a relative humidity of 85% while the devices are biased at $V_{DS} = 6V$ with a power consumption of $P < 200mW$ (according to JEDEC JESD22-A101C) [4].

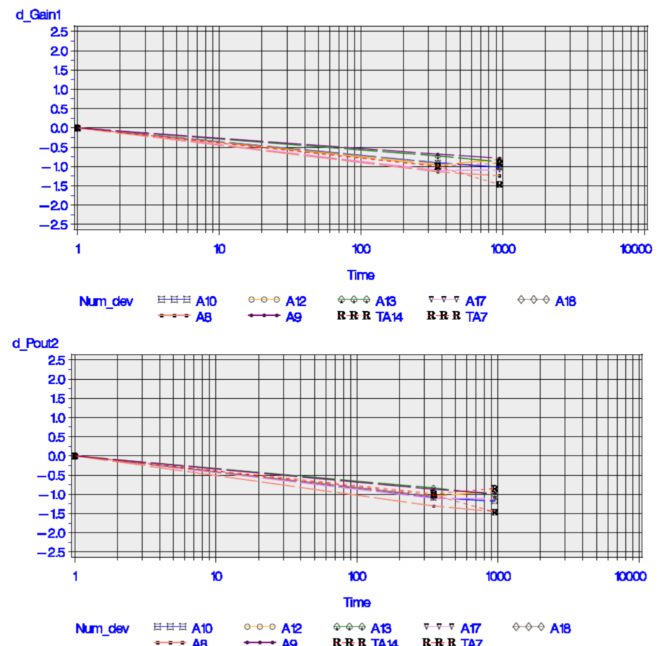

Fig. 7: RF gain and output power interim measurements of a THB test on full MMIC chips using a final BCB chip protection for 1000h.

Fig. 7 shows the interim measurement of RF gain and output power. Over the test duration of 1000h no catastrophic failures have been observed. This test demonstrates the high robustness of the process against humidity. Therefore, it can be used in non-hermetic packages like molded QFN without any humidity issue.

CONCLUSIONS

A second generation 0.15µm Ka-band GaAs power pHEMT technology has been presented. The technology is developed for the design of highly linear power amplifiers for communication system up to 40GHz. The RF power and linearity characterization has demonstrated nice state-of-the-art performance competitive with the best technologies on the market. It has been shown that the process has improved RF performance and is compatible with a final BCB chip protection without any negative impact. Furthermore, humidity tests have demonstrated a high intrinsic robustness of the process against humidity. Finally, is has been proven that UMS' PPH15X-20 process is well compatible with low-cost molded QFN packages without any performance degradation while maintaining robustness against humidity in these non-hermetic packages.

To the best of our knowledge, this is the first technology available on the world-wide market offering outstanding RF performance up to 40GHz in low-cost molded QFN plastic packages.

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ACRONYMS

HEMT: High Electron Mobility Transistor
BCB: Benzocyclobutene
PAE: Power-Added-Efficiency
IMD3: Third-Order Intermodulation Distance
PBO: Power Back-Off
PA: Power Amplifier
DEC: Dynamic Evaluation Circuit
HTOL: High-Temperature Operation Life-Test
THB: Temperature-Humidity-Biased-Test